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The dynamic resistance of YBCO coated conductor wire: Effect of DC current magnitude and applied field orientation

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Abstract: Dynamic resistance, which occurs when a HTS coated conductor carries a DC current under an AC magnetic field, can have critical implications for the design of HTS machines. Here, we report measurements of dynamic resistance in a commercially available SuperPower 4 mm-wide YBCO coated conductor, carrying a DC current under an applied AC magnetic field of arbitrary orientation. The reduced DC current, I_t/I_{c0} , ranged from 0.01 to 0.9, where I_t is the DC current level and I_{c0} is the self-field critical current of the conductor. The field angle (the angle between the magnetic field and the normal vector of the conductor wide-face) was varied between 0° and 90° at intervals of 10° . We show that the effective width of the conductor under study is $\sim 12\%$ less than the physical wire width, and we attribute this difference to edge damage of the wire during or after manufacture. We then examine the measured dynamic resistance of this wire under perpendicular applied fields at very low DC current levels. In this regime we find that the threshold field, B_{th} , of the conductor is well described by the non-linear equation of Mikitik and Brandt. However, this model consistently underestimates the threshold field at higher current levels. As such, the dynamic resistance in a coated conductor under perpendicular magnetic fields is best described using two different equations for each of the low and high DC current regimes. At low DC currents where $I_t/I_{c0} \leq 0.1$, the non-linear relationship of Mikitik and Brandt provides the closest agreement with experimental data. However, in the higher current regime where $I_t/I_{c0} \geq 0.2$, closer agreement is obtained using a simple linear expression which assumes a current-independent penetration field. We further show that for the conductor studied here, the measured dynamic resistance at different field angles is dominated by the perpendicular magnetic field component, with negligible contribution from the parallel component. Our findings now enable the dynamic resistance of a single conductor to be analytically determined for a very wide range of DC currents and at all applied field angles.

Keywords: Dynamic resistance, AC loss, YBCO, REBCO

1. Introduction

1

The current carrying performance of REBCO coated conductors has increased significantly in recent years, and has reinvigorated interest in HTS power applications utilising REBCO wires [1]. Many HTS machinery applications require an HTS wire to carry DC current whilst experiencing an applied alternating magnetic field, and in these applications dynamic resistance plays critical role [2-5]. HTS field windings in rotating machines experience an applied ripple field whilst carrying a DC current. The resulting dynamic resistance leads to heat dissipation which may impose a substantial heat load on the cryogenic system [6-8]. Dynamic resistance is also critical to the design and operation of HTS flux pumps, as it sets a limit to the maximum achievable current which such devices can deliver [9-13]. Despite these varied applications, there have been limited reports of the measurement of dynamic resistance in practical HTS coated conductors.

Experimental reports to date have focused on dynamic resistance measurements in REBCO coated conductors under perpendicular AC magnetic fields [14-17]. Several analytical equations have been developed to predict dynamic resistance in this geometry [16, 18]. If we define a thin superconductor as seen in figure 1, the dynamic resistance per unit length per cycle in a perpendicular AC magnetic field can be written as [16, 18],

$$\frac{R_{\text{dyn},\perp}}{fL} = \frac{4a}{I_{c0}} (B_{a,\perp} - B_{\text{th},\perp}), \quad (1)$$

where, a is half-width of the conductor, I_{c0} is the self-field critical current of the conductor, $B_{a,\perp}$ is the magnetic field amplitude. Recently [16, 17] we have suggested that the threshold field, $B_{\text{th},\perp}$, can be expressed as

$$B_{\text{th},\perp} = B_{p,\perp} (1-i) = \frac{4.9284 \mu_0 J_{c0} t}{\pi} (1-i), \quad (2)$$

where, i is I/I_{c0} , J_{c0} is critical current density and t is half-thickness of the conductor. $B_{p,\perp}$ is the effective penetration field which can be obtained from the maxima of the Brandt equation for AC magnetization loss [19]. It has been shown equation (2) provides good agreement with all previously-published experimental data for $i > 0.1$ [16]. However, an alternative expression for the threshold field is provided by Mikitik and Brandt, $B_{\text{th},\text{MB}}$ [18],

$$B_{\text{th},\text{MB}} = \frac{\mu_0 J_{c0} t}{2\pi} \left[\frac{1}{i} \ln \left(\frac{1+i}{1-i} \right) + \ln \left(\frac{1-i^2}{4i^2} \right) \right]. \quad (3)$$

Equations 2 and 3 diverge markedly in the low current region, and this is therefore the most sensitive region to discern between the applicability of each expression. However, until now, there have been no reported measurements of dynamic resistance for which $i < 0.1$. In addition, in actual HTS coated conductor applications, the conductors are exposed to AC magnetic fields with non-perpendicular orientations. To date, there is only one previous report of the effect of varying the applied AC field angle on the measured dynamic resistance [17], and this was limited to only three different orientations. In this work, we have addressed both of these gaps in the literature. We have measured the dynamic resistance of a 4 mm-wide YBCO coated conductor manufactured by SuperPower (SCS4050), using our newly developed measurement system [20]. We have studied a wide range of i values, including finely resolved data for $i < 0.1$. The field angle was varied from 0° to 90° at a resolution of 10° . The measured dynamic resistance was compared with the analytical expressions above, as well as to 2D finite element calculations using the T method described in chapter 3 of [21].

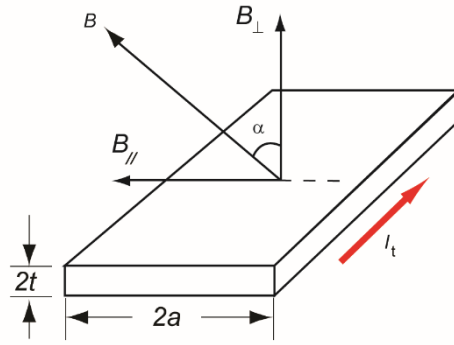


Figure 1. Schematics of field angle

2. Experimental methods

The measurement of dynamic resistance was undertaken using a similar method to that previously reported [17, 22]. A custom made AC magnet comprising two halves of racetrack coils was used to generate more than 100 mT peak amplitude at three different frequencies [20]. The 15 cm-long sample was mounted on the machined surface of a G-10 sample holder and was inserted into the magnet bore. The field angle of the sample was adjusted by rotating the sample holder whilst monitoring a one-turn pick-up coil on the flat surface of the sample holder. Current leads were arranged in order to avoid magnetic coupling as shown in figure 2(a). A DC power supply (Hewlett Packard 6682A) was used to feed DC current to the sample. The coated conductor sample was fabricated by SuperPower (SCS4050), and has a self-field I_c of 87.7 A at 77 K. Key specifications of the sample are shown in Table 1. Figure 2(b) shows the arrangement of two sets of voltage taps attached to the sample: a spiral loop was arranged around the sample for one set; whilst the second set comprised two voltage taps attached at the centre of the sample and extended from their midpoint [18, 23]. The distance between the voltage-taps was 5 cm. The voltage output from the voltage taps were measured using a DC nanovolt-meter (Hewlett Packard 34420A). The integration time (NPLC) of the nanovolt-meter was set at 10 power line cycles. Data obtained from the two sets of voltage taps was found to be almost identical, so data from the spiral loop taps has been reported throughout this work.

3. Numerical methods

The local current and magnetic flux are key factors to determine losses in HTS coated conductors. T -formulation is a finite-element electromagnetic modelling approach which was developed to help clarify how current and magnetic flux varies and affects losses within 2D superconducting films [24-26]. This method has been broadly validated and applied in quantifying losses [27-30]. However, the ac loss calculated by this approach is a total loss which includes both magnetisation loss and dynamic loss (i.e. that due to dynamic resistance). In this paper we have amended the T -model to enable these two different sources of loss to be separated, so that a calculated dynamic resistance can be compared to our experimental results. The thin-strip approximation was applied, since HTS coated conductors comprise very thin films of superconducting material with a very high aspect ratio. The governing equations are formulated using the current vector potential T instead of the current density J , which is defined by:

$$J = \nabla \times T. \quad (4)$$

The total AC loss per unit volume can be calculated from:

$$P = JE = \frac{J^2}{\sigma_{sc}} \quad (5)$$

where σ_{sc} is conductivity of the superconductor layer. When an HTS coated conductor carries a DC transport current under an AC magnetic field, the transport current I_t will occupy the central region of the superconducting film, within a distance $2ai$ in the centre of the coated conductor, leaving the rest with width $2(1-i)a$ free on both sides [16]. Consequently, the dynamic loss P_{dyn} (which is the loss due solely to dynamic resistance) may be approximately calculated by:

$$P_{dyn} = \int_{-(1-i)a}^{(1+i)a} 2JEtdy = \int_{-(1-i)a}^{(1+i)a} \frac{J^2}{\sigma_{sc}} 2tdy \quad (6)$$

where i is the ratio between transport current I_t and self-field critical current I_c and t_s is the thickness of the superconductor layer.

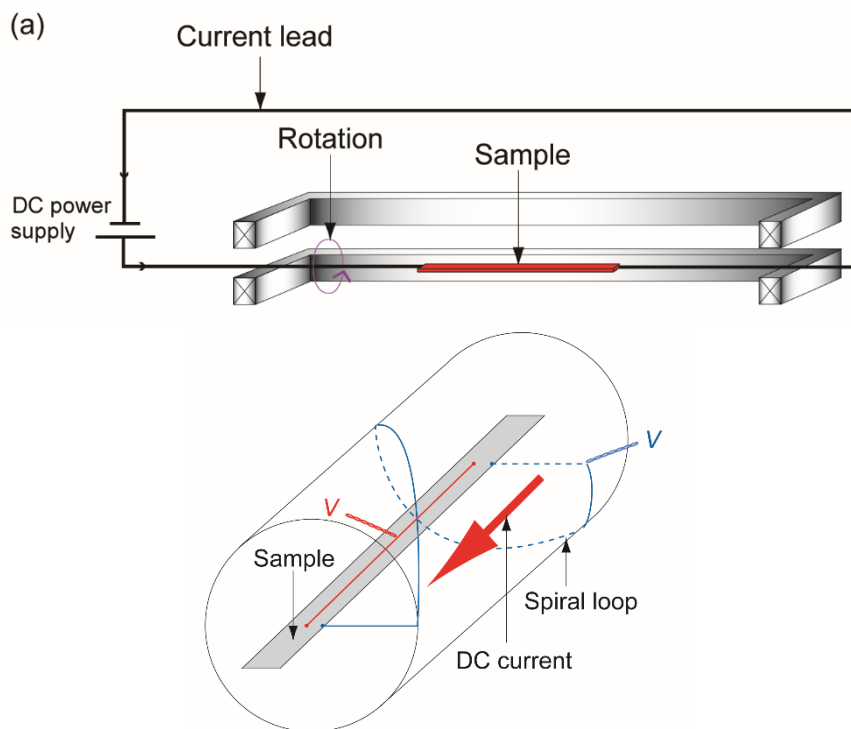


Figure 2. (a) Schematic of the measurement and (b) schematic of voltage tap arrangement

Table 1 YBCO conductor specifications

Manufacturer	SuperPower
Self-field critical current (A)	87.7 A
Sample width (mm)	4.0
Thickness of Cu stabilizer layer each side (μm)	20
Thickness of superconductor layer (μm)	1.0
Thickness of substrate (μm)	50

4. Experimental results and discussion

4.1 Measurements of R_{dyn} under alternating perpendicular magnetic fields

Figure 3 shows the measured R_{dyn} values for the sample at two different frequencies and for five different values of reduced current, $i = I/I_{c0}$. The R_{dyn} values are normalised by frequency and the sample length, and are plotted as a function of the amplitude of the applied alternating perpendicular magnetic field, $B_{a,\perp}$. In the figure, analytical values calculated from equations (1) and (2), are also plotted, along with the results of the 2D FEM computation via the T method.

For $i < 0.7$, data obtained at both frequencies falls upon a single line which intercepts the x -axis at the threshold magnetic field, $B_{\text{th},\perp}$. The near-perfect agreement between data obtained at different frequencies reflects the hysteretic nature of dynamic resistance. For $i = 0.9$, a linear correlation is observed at applied fields up to ~ 60 mT, but at higher field amplitudes an additional non-linear contribution becomes apparent. We have previously shown that this non-linear component arises due to flux flow resistance, which occurs once the applied current exceeds the field-dependent critical current of the wire for part of the cycle (i.e. once $4at \cdot J_c(B_{a,\perp}) < I_t$).

Experimental values for the threshold magnetic field, $B_{\text{th},\perp}$ were determined by fitting the linear section of the data shown in each plot in figure 3. $B_{\text{th},\perp}$ values are found to decrease with increasing DC current, as observed in previous works [16, 17]. Notably, analytically values calculated from equations (1) and (2) were observed to slightly overestimate the experimental data, showing poorer agreement than our previous report on a different SuperPower wire [16]. We observe disagreement in both $B_{\text{th},\perp}$ and the gradient, $dR_{\text{dyn}}/dB_{a,\perp}$. The numerically computed 2D-FEM results broadly agree with the analytically calculated values, but also deviate from the experimental data.

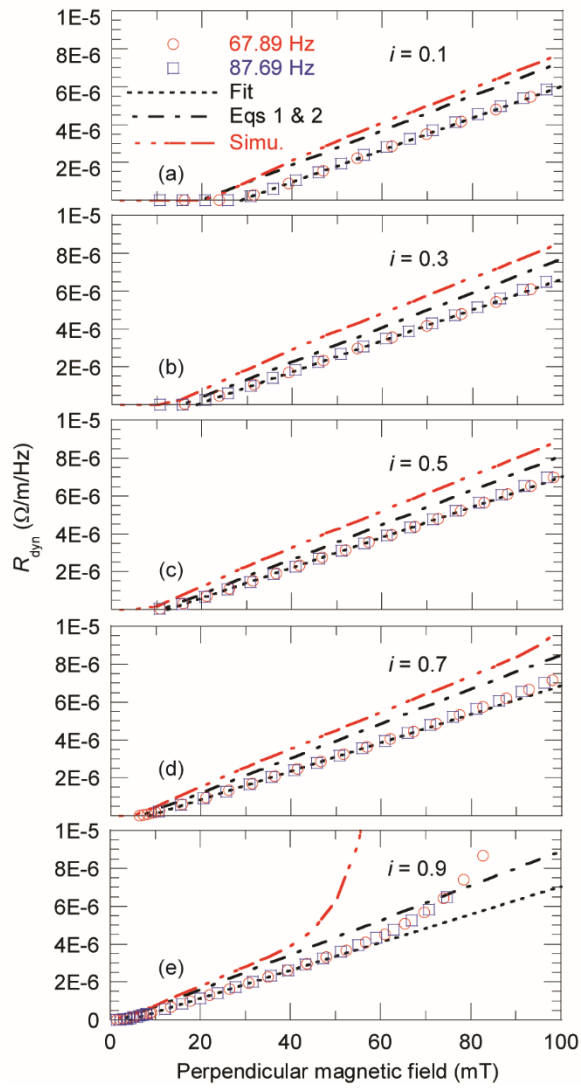


Figure 3. Measured R_{dyn} values for the sample for $i = 0.1, 0.2, 0.3, 0.5, 0.7,$ and 0.9 at 67.89 Hz and 87.69 Hz compared with values derived from the combination of equations (1) and (2), and from FEM simulation using the T -formulation.

To probe this discrepancy, figure 4 plots the gradients, $dR_{\text{dyn}}/dB_{a,\perp}$ obtained from each of the linear fits in figure 3. These are plotted as a function of I , and compared with the theoretically expected value $4a/I_{c0}$ from equation (1). The experimentally-obtained values are found to be consistent, but significantly lower than the theoretical value. We believe this difference to be due to edge effects in the coated conductor wire, resulting in an actual conductor width which is significantly smaller than the physical dimension of the wire (see figure 6). In coated conductors, the J_c values at each edge of the conductor are often found to be smaller than in the central region, due to damage caused by cutting processes [31-34]. If the true conductor width, $2a$, is actually thinner than given in Table 1, then this would explain why, in this case, both analytical and FEM calculations have under-estimated $B_{\text{th},\perp}$ and hence overestimated R_{dyn} .

Effective conductor widths are normally 90 % of the conductor width due to slitting processes [38, 39]. If we take the average value of the experimentally determined gradient values, dR_{dyn}/dB_a , the difference between the average value and the analytical value is 12.2 %, which implies an effective conductor width of 3.51 mm. This assumes a uniform lateral J_c distribution such as shown in figure 6. (Typically lateral J_c distribution are trapezoidal, however here we have assumed a simpler pattern for convenience [32-34, 36, 37].)

Further evidence for a decreased effective conductor width is found in transport AC loss measurements of the same coated conductor wire. This data is shown in figure 5, where it is plotted as a function of the amplitude of the transport current. The measured sample was cut from the same source material as the sample used in the work. I_{c0} of this short-length sample was 86.7 A. In the figure, Norris strip and ellipse models (N-s and N-e in the figure) are plotted together [35]. The significant increase of the measured results above that predicted by N-s again implies a decreased effective conductor width, as has been discussed in previous works [36, 37].

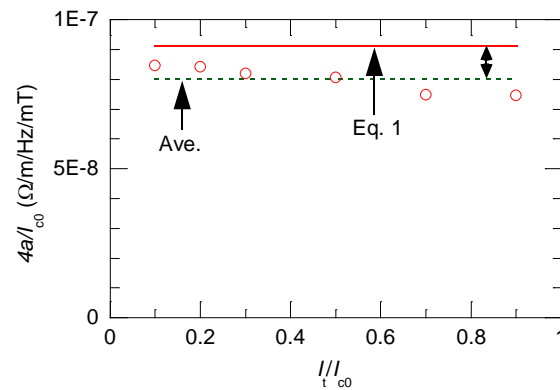


Figure 4. Comparison of the gradient of the linear fits, dR_{dyn}/dB_a vs $4a/I_{c0}$.

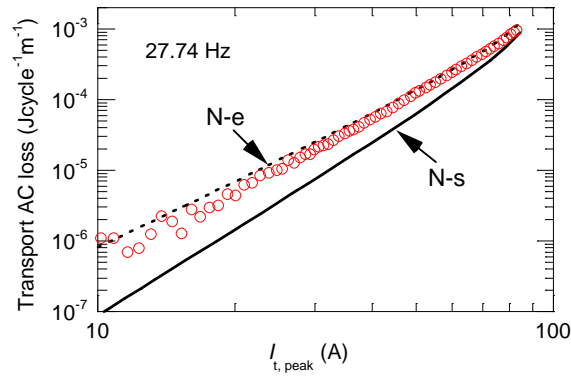


Figure 5. Transport AC loss of a YBCO conductor at 27.74 Hz. The sample is cut from the same source material as the sample used in the work. I_{c0} of the sample is 86.7 A.

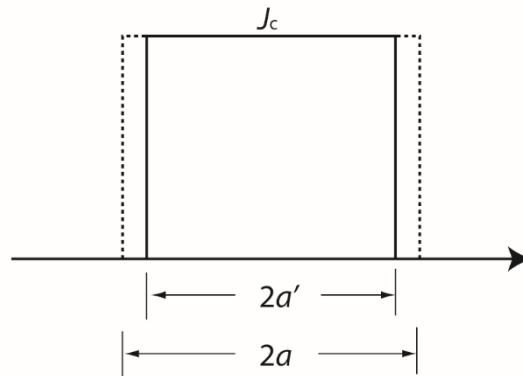


Figure 6. A schematic of assumed lateral J_c distribution and the effective conductor width. In this figure, $2a$, $2a'$ stand for the physical conductor width and effective conductor width, respectively.

In light of this we have recalculated the analytical and FEM values using an effective conductor width of, $2a = 3.51$ mm and replotted this alongside the previously shown experimental values for R_{dyn} . It is clear that the agreement between the measured data and equations (1) and (2) is now excellent for all linear regions shown. Interestingly the agreement between the measured data and 2D-FEM results is not quite as good as that between the measured data and the analytical results. Nonetheless, the agreement with the linear regime is still substantially improved over figure 3. It is interesting to note that at $i = 0.9$, the 2D-FEM model substantially over-estimates the ‘lift-off’ due to non-linear flux-flow resistance at high magnetic fields, and thus deviates substantially from the experimental data [16]. The magnitude of the flux-flow resistance is highly sensitive to the $J_c(B_{a\perp})$ behaviour of the wire, and one possible cause of the strong disagreement in this region is that the 2D-FEM calculation used here assumed a constant n value for all magnitudes of applied B . Another potential issue is that a key assumption in the FEM calculations performed here is that transport current flows only in the central region (see equation 6). This assumption may not necessarily be valid for those parts of the cycle where $I_t \gg I_c(B_a)$.

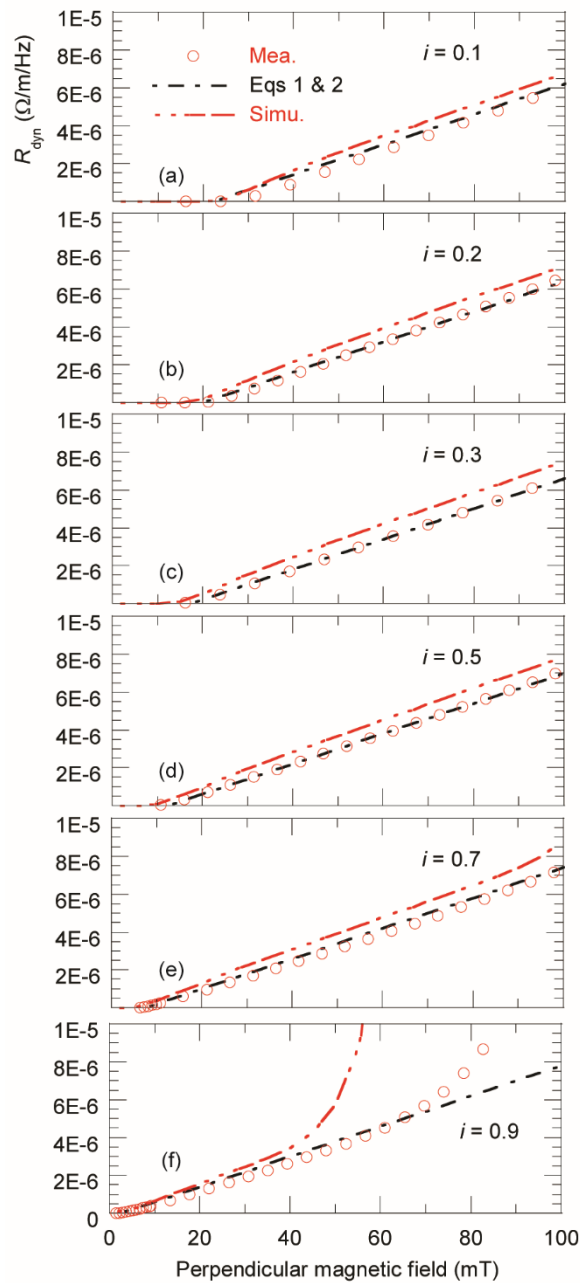


Figure 7. Measured R_{dyn} values for the sample at 67.89 Hz for $i = 0.1, 0.2, 0.3, 0.5, 0.7,$ and 0.9 compared with values derived from the combination of equations (1) and (2), and FEM simulation considering effective width of the wire.

Having established the true effective conductor width of the conductor under study, we can now examine the dynamic resistance measured at very low transport currents, where equations (2) and (3) significantly diverge from each other. Figure 8 shows the measured R_{dyn} values for $i < 0.1$, again plotted as a function of $B_{\text{a},\perp}$. Once again, analytical values calculated from equation (1) are plotted alongside the experimental data. In this case the effective conductor width $2a'$ was used, and threshold values were calculated using both equation (2) and (3). It is immediately apparent that at these low transport currents, the calculated values from equation (3)

show much better agreement with experiment than the equivalent values from equation (2). The deviation between the relative agreement of the two analytical equations becomes increasingly apparent with decreasing i , indicating that equation (2) does not appear to be valid at very low values of reduced current, where $i < 0.1$. The cause of this divergence is discussed below.

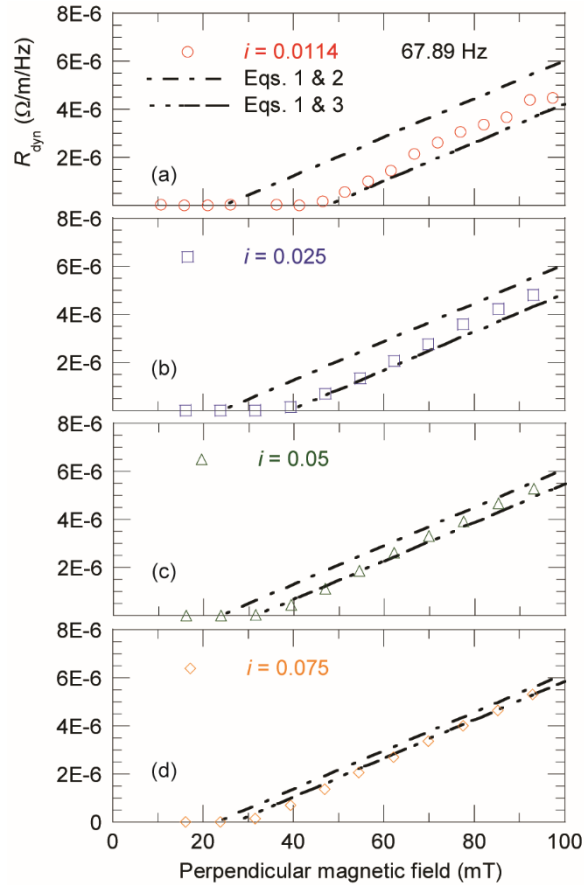


Figure 8. Measured R_{dyn} values for the sample at 67.89 Hz for $i = 0.0114, 0.025, 0.05,$ and 0.075 compared with values from the combinations of equations (1) and (2) and equations (1) and (3) considering effective wire width

Figure 9 shows the experimentally-obtained $B_{th,\perp}$ values plotted as a function of i . Also shown are calculated values from both equations (2) and (3), using both the physical and effective conductor widths, $2a$ and $2a'$, respectively. For both equations we again see that values calculated using the reduced effective conductor width provide better agreement with experiment, thus providing further confidence in this initial conclusion. It is also clearly apparent that for $i \geq 0.2$, equation (2) yields the best agreement with experiment. This is consistent with our previous work [16], in which we showed that equation (3) consistently underestimates B_{th} in this technologically-relevant regime. However, in the very low current region where $i \leq 0.1$, it is equally clear that equation (3) provides substantially better agreement with our experimental data. To understand this, we must consider the underlying assumptions behind each equation.

Equation 2 assumes that the ‘current reversal zone’ is thin and always equal to the value which occurs at the maxima of the Brandt gamma curve [16]. This is valid as long as the frozen flux region (in which the transport current flows) is significantly wider than the current reversal zone. However, at sufficiently low DC transport currents, this assumption does not hold. In this case, the current reversal zone can become compressed by the surrounding shielding currents, resulting in a consequent increase in the effective penetration field. This increase in the gradient of current-reversal leads to a reduction in the dynamic loss, but at the expense of an increase in the magnetisation loss (which is not measured in the experiments reported here). The analysis of [18] accurately describes the non-linear evolution in current and flux profile within the wire and is valid at all values of i - thus equation 3 yields more accurate values of $B_{th,\perp}$ in the very low current regime. Despite this, equation 3 consistently underestimates $B_{th,\perp}$ at higher current levels. This is because (3) assumes that the net flux passing the centre of the current reversal zone must always interact dissipatively with the transport current. In fact, at fields just below $B_{th,\perp}$, flux can exit from the same side of the zone as it entered, leading to zero net dissipation over the full cycle. Equation (2) accounts for this by assuming that all flux is ejected non-dissipatively until the current reversal zone is compressed to the same width as occurs at the maxima of the AC magnetisation loss (the Brandt gamma curve) [16]. Empirically, this delivers better agreement with the experimental data for $i \geq 0.2$, whilst also being advantageously simple to apply.

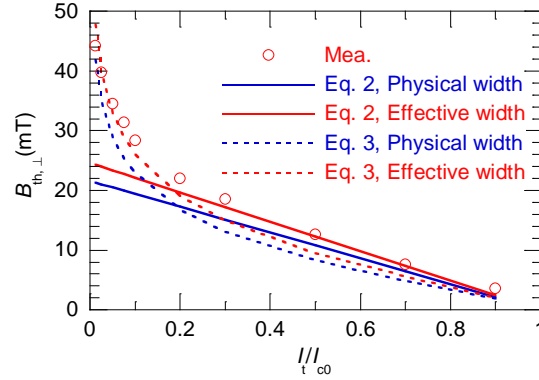


Figure 9. Comparison of $B_{th,\perp}$ values from experiment and equations (2) and (3) using physical and effective wire width.

4.2 Angular dependence of R_{dyn} under non-perpendicular alternating magnetic fields

Figure 10 shows experimentally-measured R_{dyn} values obtained at ten different angles of applied magnetic field, plotted versus the field amplitude, B_a . As before, linear fits of the experimental data are shown, and used to obtain the threshold field values, B_{th} from the x -axis intercepts. For a given value of DC current i , B_{th}

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2 decreases with decreasing field angle α , whilst the gradient of the fitted linear lines increases with decreasing
3 α .

4 At $i = 0.9$ and for low field angle we once again observe a non-linear contribution to R_{dyn} at high field
5 amplitudes. Similarly to the perpendicular field data, we attribute this to flux flow resistance arising once
6 $I_c(B_a) < I_t$ [16]. It is apparent that this non-linear contribution decreases rapidly with increasing field angle
7 such that 'lift-off' from the linear trend is negligible once $\alpha \geq 50^\circ$. This reflects the lower sensitivity of I_c to
8 non-perpendicular magnetic fields.

9 We do not observe any measurable values of R_{dyn} under parallel magnetic field ($\alpha = 90^\circ$) at any of the
10 transport currents studied, except for an extremely small onset at $i = 0.9$ and $B_a > 80$ mT. This implies that
11 parallel magnetic field plays almost no role in the dynamic resistance of the coated conductor, which is
12 entirely different to the case for BSCCO wires [4, 40]. The perpendicular component of the applied magnetic
13 field can be expressed as $B_{a,\perp} = B_a \cos\alpha$, meaning that the threshold value at angle α , can similarly be
14 expressed as $B_{\text{th}}(\alpha) = B_{\text{th},\perp} / (\cos\alpha)$. As such, we expect $B_{\text{th}}(\alpha)$ to increase with increasing α , and this is
15 consistent with the trend observed in figure 10. Similarly, the gradient $dR_{\text{dyn}}/dB_{a,\perp}$ in equation (1) now
16 becomes $dR_{\text{dyn}}/dB_a = (4a' \cdot \cos\alpha) / I_{c0}$, which is also consistent with the observed trend in this figure.
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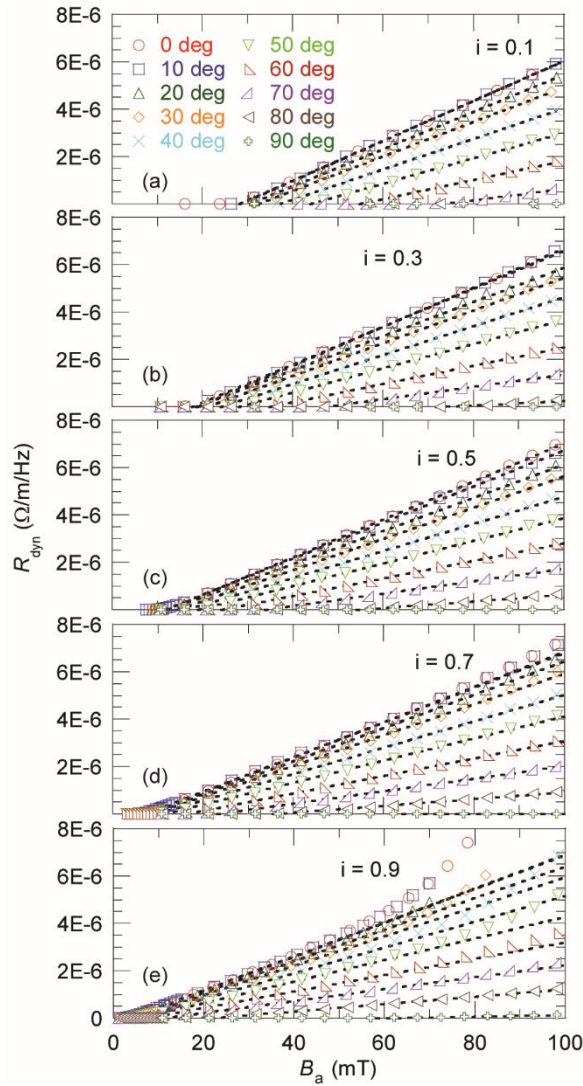


Figure 10. Measured R_{dyn} values at various field angles plotted as a function of applied magnetic field amplitude at 67.89 Hz (a) $i = 0.1$, (b) $i = 0.3$, (c) $i = 0.5$, (d) $i = 0.7$, and (e) $i = 0.9$.

Figure 11 shows fitted values of $B_{\text{th}}(\alpha)$ obtained from the experimental data at various angles, and plotted as a function of the reduced current, i . The dotted lines show calculated values using equation (2) for $i \geq 0.2$ (taking $B_{\text{th}} = B_{\text{th},\perp}/(\cos \alpha)$), and using equation (3) for $i < 0.2$ (taking $B_{\text{th}} = B_{\text{th},\text{MB}}/(\cos \alpha)$). Excellent agreement is obtained between the experimental and calculated values of B_{th} for all $\alpha < 80^\circ$, indicating that the threshold field is determined solely by the perpendicular field component of the applied field. The small perpendicular field component at $\alpha = 80^\circ$ results in a large threshold field (80 mT at $i=0.3$). This meant that there were few data points available to perform linear fits to the experimental data (see figures 10 (a), (b) and (c)). As a result, the fitted values obtained at this field angle must be treated with a little caution, and this is the likely cause of the divergence between calculated and experimental values observed at this nearly-parallel field angle.

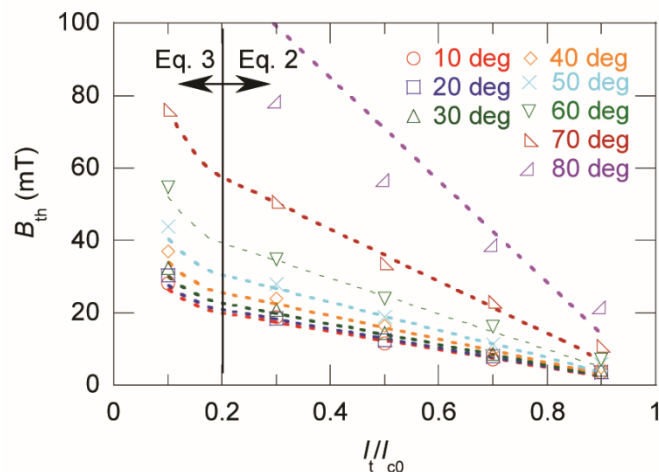


Figure 11. Experimentally-determined values for the threshold magnetic field, B_{th} obtained for various field angles, α and plotted versus the reduced transport current, $i = I_t/I_{c0}$. Dotted lines show calculated values using $B_{th} = B_{th,\perp}/(\cos \alpha)$ for $i \geq 0.2$ (from equation 2), and $B_{th} = B_{th,MB}/(\cos \alpha)$ for $i < 0.2$ (from equation 3).

Finally, in figure 12, the measured R_{dyn} values (shown previously in figure 10) are now re-plotted as a function of the perpendicular component of the applied magnetic field, $B_{a,\perp} = B_a \cos \alpha$. The entire experimental dataset is shown, together with the values calculated from equations (1) and (2). The R_{dyn} values measured at each different field angle now collapse onto a common curve, and show excellent agreement with the analytically calculated values for all linear regions shown. Taken together, the results in figures 11, and 12 provide compelling evidence that (for all $\alpha \leq 80^\circ$), the dynamic resistance of this coated conductor wire is determined solely by the perpendicular component of the applied magnetic field.

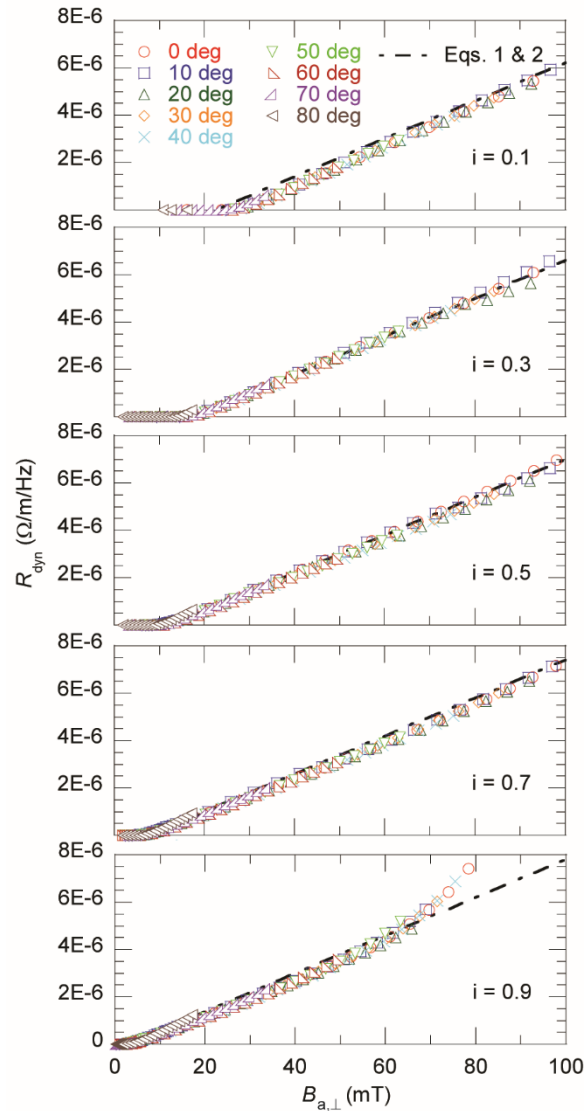


Figure 12. Measured R_{dyn} values at various field angles plotted as a function of perpendicular magnetic field component, $B_{a,\perp} = B_a \cos \alpha$. Data obtained at 67.89 Hz for (a) $i = 0.1$, (b) $i = 0.3$, (c) $i = 0.5$, (d) $i = 0.7$, and (e) $i = 0.9$.

5. Conclusion

In this work, we have measured dynamic resistance in a 4 mm-wide YBCO coated conductor carrying DC currents under AC magnetic fields incident at various different field angles. Measurements were carried out using a wide range of DC currents from $0.0114 \leq i \leq 0.9$, and the applied field angle was changed from zero to 90° with a resolution of 10° .

When the conductor width was taken to be the measured physical dimension of the wire, we found that experimentally-measured R_{dyn} values did not agree well with calculated analytical values, nor with the numerical 2D-FEM results computed using the T – method, nor with transport AC loss measurements of the same wire. We have attributed this difference to a non-uniform critical current distribution at the edges of the coated conductor wire, which results in a reduced effective conductor width. To account for this, we have calculated an equivalent conductor width from the measured gradient, dR_{dyn}/dB_m for all measured values at $i > 0.1$. This value was confirmed through independent measurements of the AC transport loss of this wire. Using this effective conductor width, we then obtain substantially better agreement between experiment and both our analytical and 2D-FEM calculations.

We have compared our experimental data with two different analytical expressions which have been proposed to describe the threshold field under perpendicular fields. Interestingly, we find that the relative agreement between these two equations and experiment changes depending on the magnitude of the reduced current, i . For $i \geq 0.2$, we find that the simple linear expression within equation (2) provides best agreement with experiment, which is consistent with previously reported results in [16]. This high current regime is technologically important, as it corresponds to situations in which most of current-carrying capacity of the wire is being employed. However, at very low DC current levels ($i \leq 0.1$) we find that experimental values of B_{th} increase sharply with decreasing i and are thus best described by the equation of Mikitik and Brandt (equation (3)). We believe that our results represent the first robust experimental confirmation of the work presented in [18] for the low DC current regime.

We have also measured the dynamic resistance which occurs at non-perpendicular applied field angles. We have found that this is dominated solely by the perpendicular field component. This means that R_{dyn} can be simply calculated using solely the expressions which describe the perpendicular field behaviour (unless the field angle is close or equal to 90°). The values of R_{dyn} measured under parallel magnetic fields are more than three orders of magnitude smaller than those observed under perpendicular field, and this renders these values negligible in most practical applications. From an experimental standpoint, such extremely small DC signals are challenging to measure, and the measured value is highly sensitive to minor misalignment of the sample in the applied field. As such, precise determination of the parallel field behaviour of coated conductor wires remains a largely unexplored avenue for further future work.

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